Understanding the Vulnerability of a Mixed-Source Microgrid to Malicious Control of an Active Load

SPADES Workshop 2nd December 2020



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- Previous work examined malicious control for synchronous machine dominated bulk power system [1-3]
- ▶ We focus on mixed-source microgrid for two main reasons:
 - the relative size of individual loads may make it easier for an adversary to gain sufficient controllability
 - these systems are among the first to achieve very high penetration of converter-based generation
- We examine the use of eigenstructure assignment, eigenvalue and eigenvector design, to develop a feedback controller to destabilize a vulnerable mode of the system
 - Will consider two distinct attacks as well as two different levels of adversarial control within each attack



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Microgrid Model



- Nominal loading of 1 p.u.
- Synchronous machine (SG), Active Load (AL), and Grid-following converter (Gf)
 - System has 80% penetration of Gf with AL 5% nominal active power

State space
$$x \in \mathbb{R}^{46}$$

► $x^{SG} \in \mathbb{R}^{13}$, $x^{AL} \in \mathbb{R}^{12}$, $x^{Gf} \in \mathbb{R}^{15}$, $x^{PI} \in \mathbb{R}^2$, $x^{Net} \in \mathbb{R}^4$





AL consists three different PI control loops

- 1. a phase-locked loop (PLL) for alignment of internal and network synchronously-rotating (*dq*) reference frame (SRF)
- 2. an outer-loop voltage controller for maintaining a constant DC voltage across the DC-link capacitor
- 3. and an inner-loop current controller for current tracking



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Adversarial Level of Access



- For each attack we consider two different levels of adversarial control
 - 1. For setpoint control, the adversary only has the ability to change the DC voltage setpoint. In this case, we have that the input $u \in \mathbb{R}$.
 - 2. For full control, the adversary has full access to the inner control loops and, therefore, $u \in \mathbb{R}^5$.
- The dimension of the input will determine the flexibility in designing the eigenvector(s) for the unstable eigenvalue(s)



We linearize our set of non-linear equations about an operating point to give us

$$\Delta \dot{\boldsymbol{x}} = \boldsymbol{A} \Delta \boldsymbol{x} + \boldsymbol{B} \Delta \boldsymbol{u}, \tag{1}$$

We want to design a linear state feedback controller for the AL to force a stable mode of A to become unstable wile minimizing the participation of the AL in this new unstable mode. The participation of state *i* in mode *j* is defined as

$$p_{ij} = \frac{w_{ij}v_{ji}}{\boldsymbol{w}_j^T \boldsymbol{v}_j},\tag{2}$$

were w and v are the left and right eigenvectors respectively. We will optimize the elements of the eigenvector \hat{v} , corresponding to the desired unstable eigenvalue $\hat{\lambda}$, to minimize the participation of the AL in the unstable mode[2]. (See Appendix)

Destabilizing Modes



Attack 1: $\lambda = -0.29 \pm 1.279j$ → $\hat{\lambda} = 0.29 \pm 1.279j$ Attack 2: $\lambda = -0.45 \pm 16.16j$ → $\hat{\lambda} = 0.45 \pm 16.16j$



Participation Factors for Unstable Mode



- Attack 1: SG mechanical power, p_m and angular frequency, ω_s are the dominant states
- Attack 2: The states for the Gf active power controller and PLL are dominant states



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Attack 1: SG Behavior



- Attack 1 is largely SG based and closely resembles attack reported in prior work
- Similar behavior for both levels of adversarial access considered



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Attack 1: AL Behavior



 Full control allows adversary to independently control active and reactive power

Setpoint control can only affect the active power demand

- For full control, AL appears capacitive when maximizing active power demand
 - Raising/lowering the voltage the voltage dependency of the constant impedance load



Attack 2: Gf Behavior



- Attack 2 is largely Gf based and is first identified in this work
- Response very different for each level of access
 - Oscillation grows at a much faster rate for case of full control



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Attack 2: AL Behavior



- Similar to previous attack, setpoint control only controls active power
- In the case, full control exerts less control effort to induce instability
 - Both active power and reactive power oscillation magnitude ≈2-3%



Hardening the System



- What if we replace synchronous machine with a grid-forming inverter?
- Initial analysis suggests we increase the stability boundary and harden the system against the particular attack vector considered

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- Consideration of more sophisticated adversarial attacks, e.g. non-linear controllers characterized by neural networks
- Impact of microgrid composition, i.e. continual examination of grid following vs grid forming, different levels of synchronous generation
- Examination of adaptive paramertization of controllers to increase stability margin



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Once we have identified a candidate mode to be destabilized with a desired eigenvalue $\hat{\lambda}$, we begin by constructing the corresponding Hautus matrix $\boldsymbol{S}_{\hat{\lambda}}$ given by

$$\mathbf{S}_{\hat{\lambda}} = \begin{bmatrix} (\hat{\lambda} \mathbf{I} - \mathbf{A}) & \mathbf{B} \end{bmatrix}$$
(3)

where ${\it I}$ is the identity matrix. We then determine the matrix ${\it K}_{\hat{\lambda}}$ of the form

$$\boldsymbol{K}_{\hat{\lambda}} = \begin{bmatrix} \boldsymbol{N}_{\hat{\lambda}} \\ \boldsymbol{M}_{\hat{\lambda}} \end{bmatrix}, \qquad (4)$$

whose columns form a basis for nullspace of $S_{\hat{\lambda}}$. $N_{\hat{\lambda}} \in \mathbb{R}^{n \times m}$ and $M_{\hat{\lambda}} \in \mathbb{R}^{m \times m}$.



The eigenvector $\hat{\boldsymbol{v}}$, is then expressed as

$$\hat{\boldsymbol{\nu}} = \boldsymbol{N}_{\hat{\lambda}} \boldsymbol{k}$$
 (5)

for some $\mathbf{k} \in \mathbb{R}^{m \times 1}$. We let $\mathbf{N}_{\hat{\lambda}\mathcal{T}}$ and $\mathbf{N}_{\hat{\lambda}C}$ denote the rows of $\mathbf{N}_{\hat{\lambda}}$ whose indices correspond to the states of the target and control group, respectively. We then seek to determine the optimal design vector \mathbf{k}^* for maximizing the ratio of ℓ_2 -norm of the eigenvector entries corresponding to the target states and the ℓ_2 -norm of the eigenvector entries corresponding to the control group.

$$\max_{\boldsymbol{k}} \quad \frac{\boldsymbol{k}'[\boldsymbol{N}_{\hat{\lambda}T}]'\boldsymbol{N}_{\hat{\lambda}T}\boldsymbol{k}}{\boldsymbol{k}'[\boldsymbol{N}_{\hat{\lambda}C}]'\boldsymbol{N}_{\hat{\lambda}C}\boldsymbol{k}} \tag{6}$$

s.t.
$$\boldsymbol{k}'\boldsymbol{k} = 1,$$

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Defining the matrices \boldsymbol{G} and \boldsymbol{H} as

$$\boldsymbol{G} = [\boldsymbol{N}_{\hat{\lambda}T}]' \boldsymbol{N}_{\hat{\lambda}T} \quad \boldsymbol{H} = [\boldsymbol{N}_{\hat{\lambda}C}]' \boldsymbol{N}_{\hat{\lambda}C}, \tag{7}$$

we rewrite this optimization as

$$\max_{\boldsymbol{\nu}} \frac{\boldsymbol{\nu}'(\boldsymbol{H}^{-1/2})^T \boldsymbol{G} \boldsymbol{H}^{-1/2} \boldsymbol{\nu}}{\boldsymbol{\nu}' \boldsymbol{\nu}}.$$
 (8)

The optimal design vector \mathbf{k}^{\star} is constructed using the eigenvector ν_{max} corresponding to the largest eigenvalue of (8), and is given by

$$\boldsymbol{k}^{\star} = \boldsymbol{H}^{-1/2} \boldsymbol{\nu}_{max}. \tag{9}$$

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We then construct the feedback matrix F follows. Let us define

$$\hat{\boldsymbol{w}} = \boldsymbol{M}_{\hat{\lambda}} \boldsymbol{k}^{\star}, \ \hat{\boldsymbol{\nu}} = \boldsymbol{N}_{\hat{\lambda}} \boldsymbol{k}^{\star}, \tag{10}$$

and construct the real matrices \boldsymbol{W} and \boldsymbol{V} of the form

$$\boldsymbol{W} = [Re\{\hat{\boldsymbol{w}}\} \, Im\{\hat{\boldsymbol{w}}\} \, 0 \, \dots \, 0], \tag{11a}$$

$$\boldsymbol{V} = [Re\{\hat{\boldsymbol{v}}\} Im\{\hat{\boldsymbol{v}}\} Re\{\boldsymbol{v}_3\} Im\{\boldsymbol{v}_3\} \dots \boldsymbol{v}_{n-1} \boldsymbol{v}_n], \quad (11b)$$

where $[\mathbf{v}_3, \dots, \mathbf{v}_{n-1}, \mathbf{v}_n]$ are the remaining original eigenvectors from the state-space matrix \mathbf{A} given in (1). The feedback matrix \mathbf{F} is then given by

$$F = \mathbf{W}\mathbf{V}^{-1}.$$
 (12)

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